

TENTATIVE

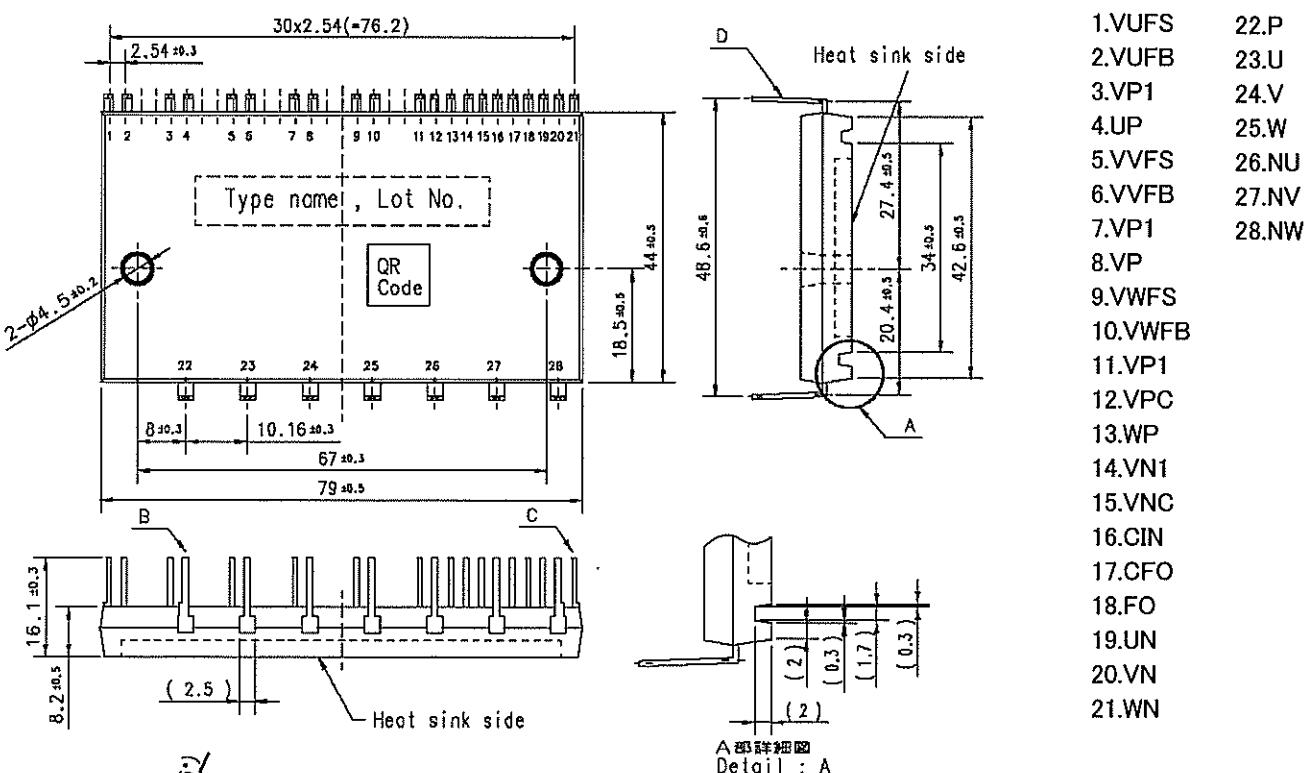
Pre.	Silou, M. Saito, M. Sato, T. Inoue	Rev.	A
Apr.	2004 Nov. 30, Nov. '04		

**Applications :** AC100~200V three-phase motor variable speed inverter drive.**Integrated Power Functions :**

600V/20A low-loss 5<sup>th</sup> generation planar gate IGBT inverter bridge with N-side open emitter structure for DC-to-AC power conversion

**Integrated drive, protection and system control functions :**

- For upper-leg IGBTs : Drive circuit, High voltage isolated high-speed level shifting, Control supply under-voltage (UV) protection.
- For lower-leg IGBTs : Drive circuit, Control supply under-voltage protection (UV), Short circuit protection (SC).
- Fault signaling : Corresponding to an SC fault (Lower-side IGBT) or a UV fault (Lower-side supply).
- Input interface : 5V line CMOS/TTL compatible, Schmitt Trigger receiver circuit (High Active).

**Fig. 1 Package Outlines**

Note 1: Package outlines are subjected to change in the development.

2: All lead terminals are treated by Pb-free solder (ingredient: Sn-Cu) plating.

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Transfer-Mold Type  
Insulated TypeMaximum Ratings ( $T_j=25^\circ\text{C}$ , unless otherwise noted) :

Inverter Part:

Item	Symbol	Condition	Rating	Unit
Supply voltage	$V_{CC}$	Applied between P-NU,NV,NW	450	V
Supply voltage (surge)	$V_{CC(\text{surge})}$	Applied between P-NU,NV,NW	500	V
Collector-emitter voltage	$V_{CES}$		600	V
Each IGBT collector current	$\pm I_C$	$T_c=25^\circ\text{C}$	20	A
Each IGBT collector current (peak)	$\pm I_{CP}$	$T_c=25^\circ\text{C}$ , less than 1ms	40	A
Collector dissipation	$P_C$	$T_c=25^\circ\text{C}$ , per 1 chip	52.6	W
Junction temperature	$T_j$	(Note 1)	-20~+125	$^\circ\text{C}$

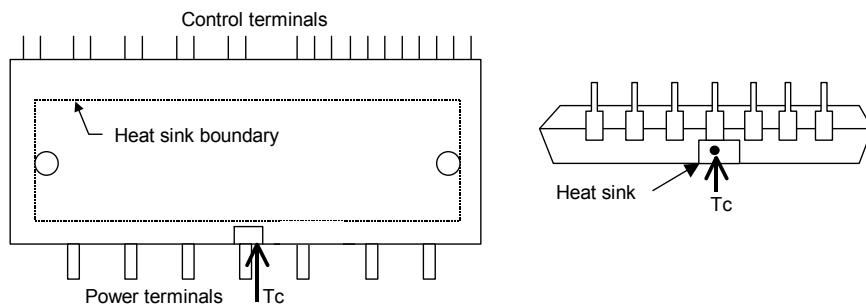
(Note 1) The maximum junction temperature rating of the power chips integrated within the DIP-IPM is  $150^\circ\text{C}$ (@ $T_c \leq 100^\circ\text{C}$ ). However, in order to ensure safe operation of the DIP-IPM, the average junction temperature should be limited to  $T_{j(\text{ave})} \leq 125^\circ\text{C}$ (@ $T_c \leq 100^\circ\text{C}$ ).

Control (Protection) Part

Item	Symbol	Condition	Rating	Unit
Control supply voltage	$V_D$	Applied between $V_{P1}-V_{PC}, V_{N1}-V_{NC}$	20	V
Control supply voltage	$V_{DB}$	Applied between $V_{UFB}-V_{UFS}, V_{VFb}-V_{VFS}, V_{WFb}-V_{WFS}$	20	V
Input voltage	$V_{IN}$	Applied between $U_P, V_P, W_P-V_{PC}, U_N, V_N, W_N-V_{NC}$	-0.5~ $V_D+0.5$	V
Fault output supply voltage	$V_{FO}$	Applied between $Fo-V_{NC}$	-0.5~ $V_D+0.5$	V
Fault output current	$I_{FO}$	Sink current at $Fo$ terminal	1	mA
Current sensing input voltage	$V_{SC}$	Applied between $CIN-V_{NC}$	-0.5~ $V_D+0.5$	V

Total System

Item	Symbol	Condition	Rating	Unit
Self protection supply voltage limit (short circuit protection capability)	$V_{CC(\text{PROT})}$	$V_D=13.5\sim16.5\text{V}$ , Inverter part $T_j=125^\circ\text{C}$ , non-repetitive less than 2 $\mu\text{s}$	400	V
Module case operation temperature	$T_c$	(Note 2)	-20~+100	$^\circ\text{C}$
Storage temperature	$T_{stg}$		-40~+125	$^\circ\text{C}$
Isolation voltage	$V_{iso}$	60Hz, Sinusoidal, AC 1 minutes, connection pins to heat-sink plate	2500	Vrms

(Note 2)  $T_c$  measurement position

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## Thermal Resistance :

Item	Symbol	Condition	Min.	Typ.	Max.	Unit
Junction to case thermal resistance	$R_{th(j-c)Q}$	Inverter IGBT part (per 1/6 module)	—	—	1.90	$^{\circ}\text{C} / \text{W}$
	$R_{th(j-c)F}$	Inverter FWD part (per 1/6 module)	—	—	2.85	
Contact thermal resistance (Note 3)	$R_{th(c-f)}$	Between case and fin with grease applied (per 1 module)	—	—	0.047	

(Note 3) Grease with good thermal conductivity and long-term endurance should be applied evenly with about  $+100\mu\text{m} \sim +200\mu\text{m}$  on the contacting surface of DIP-IPM and heat-sink.

Electrical Characteristics (  $T_j=25^{\circ}\text{C}$ , unless otherwise noted ):

## Inverter Part

Item	Symbol	Condition	Min.	Typ.	Max.	Unit
Collector-emitter saturation voltage	$V_{CE(\text{sat})}$	$V_D=V_{DB}=15\text{V}$	—	1.60	2.10	$\text{V}$
		$V_{IN}=5\text{V}, I_C=20\text{A}, T_j=125^{\circ}\text{C}$	—	1.70	2.20	
FWD forward voltage	$V_{EC}$	$T_j=25^{\circ}\text{C}, V_{IN}=0\text{V}, -I_C=20\text{A}$	—	1.50	2.00	$\text{V}$
Switching time	$t_{on}$	$V_{CC}=300\text{V}, V_D=V_{DB}=15\text{V}$ $V_{IN}=5\Rightarrow 0\text{V}, I_C=20\text{A}$ $T_j=125^{\circ}\text{C}$ Inductive load (upper-lower arm)	0.70	1.30	1.90	$\mu\text{s}$
	$t_{rr}$		—	0.30	—	
	$t_{c(on)}$		—	0.40	0.60	
	$t_{off}$		—	1.60	2.20	
	$t_{c(off)}$		—	0.50	0.80	
Collector-emitter cut-off current	$I_{CES}$	$V_{CE}=V_{CES}$	$T_j=25^{\circ}\text{C}$	—	—	1
			$T_j=125^{\circ}\text{C}$	—	—	10

## Control (Protection) Part:

Item	Symbol	Condition	Min.	Typ.	Max.	Unit	
Circuit current	$I_D$	$V_D=V_{DB}=15\text{V}$	Total of $V_{P1}-V_{PC}, V_{N1}-V_{NC}$	—	—	7.00	
		$V_{IN}=5\text{V}$	$V_{UFB}-V_{UFS}, V_{VFB}-V_{VFS}, V_{WFB}-V_{WFS}$	—	—	0.55	
		$V_D=V_{DB}=15\text{V}$	Total of $V_{P1}-V_{PC}, V_{N1}-V_{NC}$	—	—	7.00	
		$V_{IN}=0\text{V}$	$V_{UFB}-V_{UFS}, V_{VFB}-V_{VFS}, V_{WFB}-V_{WFS}$	—	—	0.55	
Fo output voltage	$V_{FOH}$	$V_{sc}=0\text{V}$ , Fo circuit: $10\text{k}\Omega$ to $5\text{V}$ pull-up	4.9	—	—	$\text{V}$	
	$V_{FOL}$	$V_{sc}=1\text{V}, I_{FO}=1\text{mA}$	—	—	0.95		
Input current	$I_{IN}$	$V_{IN}=5\text{V}$	1.0	1.5	2.0	$\text{mA}$	
short circuit trip level	$V_{SC(\text{ref})}$	$T_j=25^{\circ}\text{C}, V_D=15\text{V}$	(Note 4)	0.43	0.48	0.53	
Control supply under-voltage protection	$UV_{DBt}$	$T_j \leq 125^{\circ}\text{C}$		10.0	—	12.0	
	$UV_{DBr}$			10.5	—	12.5	
	$UV_{Dt}$			10.3	—	12.5	
	$UV_{Dr}$			10.8	—	13.0	
Fault output pulse width	$t_{FO}$	$C_{FO}=22\text{nF}$	(Note 5)	1.0	1.8	—	
ON threshold voltage	$V_{th(on)}$	Applied between $U_P, V_P, W_P-V_{PC}$ ,		2.1	2.3	2.6	
OFF threshold voltage	$V_{th(off)}$	$U_N, V_N, W_N-V_{NC}$		0.8	1.4	2.1	

(Note 4) Short circuit protection functions only for the N-side IGBTs. Please select the external shunt resistance such that the SC trip-level is less than 1.7 times of the rated current.

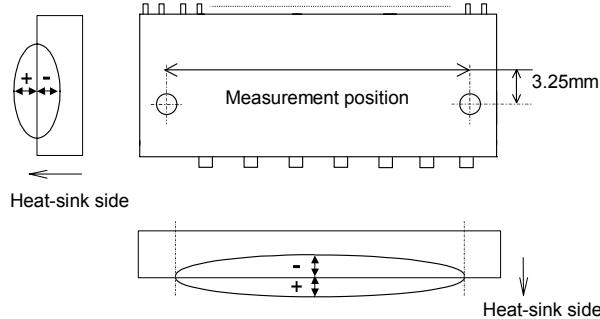
(Note 5) Fault signal is output when the lower arms short circuit or control supply under-voltage protection happens. The fault output pulse-width  $t_{FO}$  depends on the capacitance value of  $C_{FO}$  ( $C_{FO}=12.2 \times 10^{-6} \times t_{FO} [\text{F}]$ )

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## Mechanical Characteristics and Ratings:

Item	Condition		Min.	Typ.	Max.	Unit
Mounting torque	Mounting screw: (M4)	Recommended: 1.18N·m	0.98	—	1.47	N·m
Weight			—	77	—	g
Heat-sink flatness		(Note 6)	-50	—	100	μm

(Note 6)



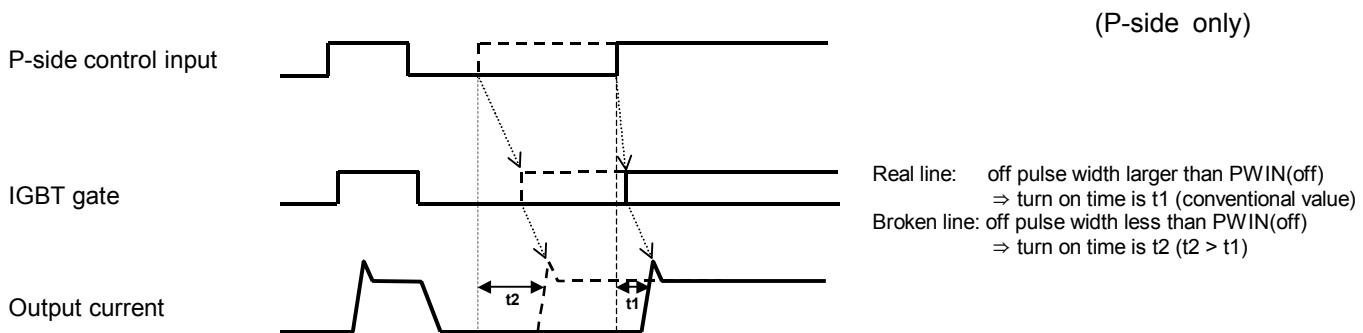
## Recommended Operation Conditions:

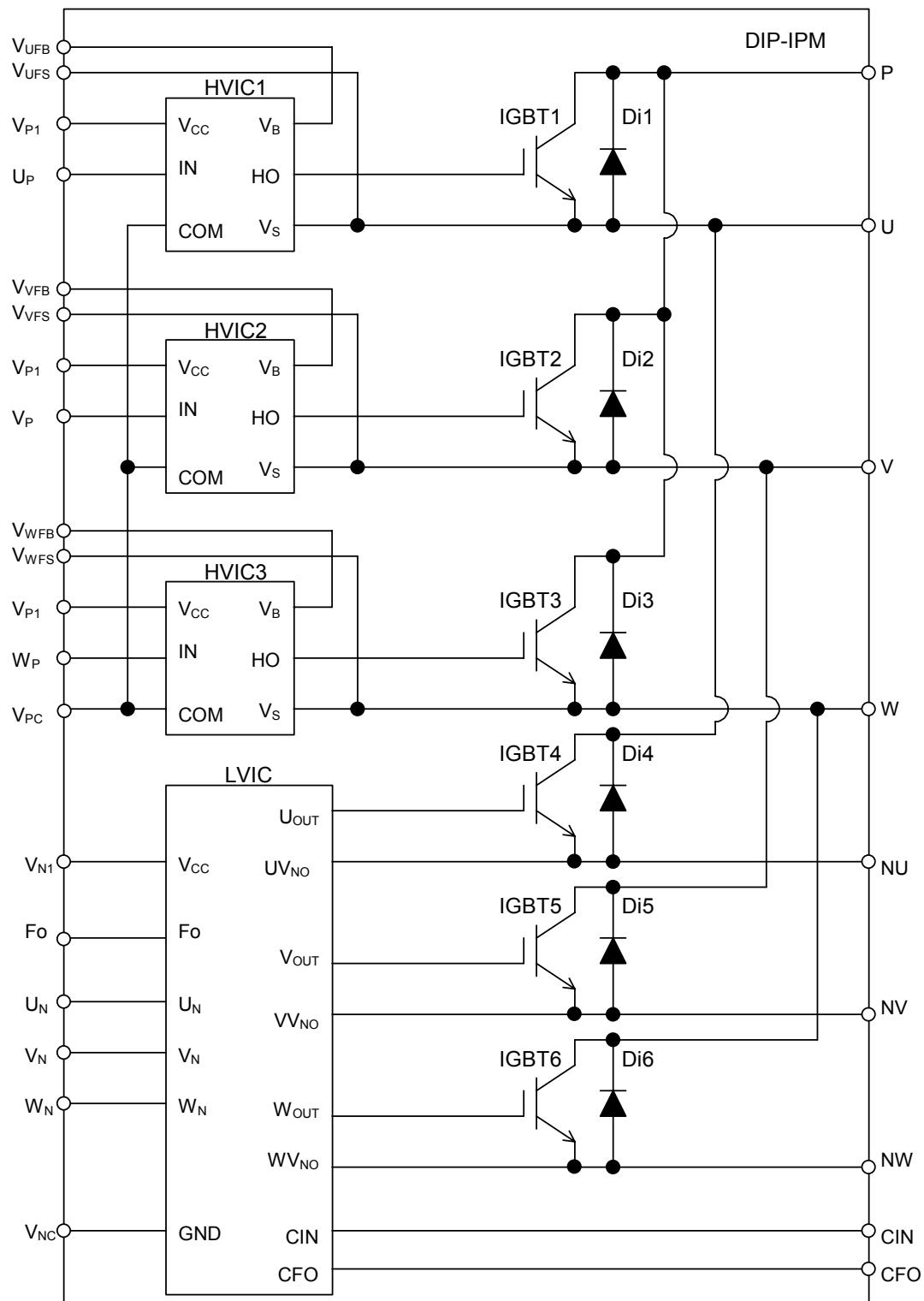
Item	Symbol	Condition	Recommended			Unit
			Min.	Typ.	Max.	
Supply voltage	$V_{CC}$	Applied between P-NU,NV,NW	0	300	400	V
Control supply voltage	$V_D$	Applied between $V_{P1}-V_{PC}, V_{N1}-V_{NC}$	13.5	15.0	16.5	V
Control supply voltage	$V_{DB}$	Applied between $V_{UFB}-V_{UFS}, V_{VFB}-V_{VFS}, V_{WFB}-V_{WFS}$	13.0	15.0	18.5	V
Control supply variation	$\Delta V_D, \Delta V_{DB}$		-1	—	+1	V/μs
Arm-shoot-through blocking time	$t_{dead}$	For each input signal, $T_c \leq 100^\circ C$	2.0	—	—	μs
PWM input frequency	$f_{PWM}$	$T_c \leq 100^\circ C, T_j \leq 125^\circ C$	—	—	20	kHz
Minimum input pulse width	PWIN(on)	(Note 7)			0.3	—
	PWIN(off)	$200 \leq V_{CC} \leq 350V, 13.5 \leq V_D \leq 16.5V, 13.0 \leq V_{DB} \leq 18.5V, -20 \leq T_f \leq 100^\circ C, N$ line wiring inductance less than $10nH$ (Note 8)	$I_C \leq 20A$	1.4	—	μs
$V_{NC}$ variation	$V_{NC}$	Potential difference between $V_{NC}-NU,NV,NW$ including surge voltage	-5.0	—	+5.0	V

(Note 7) DIP-IPM might make no response to the input on signal with pulse width less than PWIN(on).

(Note 8) DIP-IPM might make no response to the input off signal with pulse width less than PWIN(off), or P-side only the turn on time becomes long as shown in Fig.2. However, off-latch will not happen for next input on signal in this case. For the wiring inductance of N line, please refer to Fig.6.

Fig.2 Output behavior under short input off signal with pulse width less than PWIN(off)



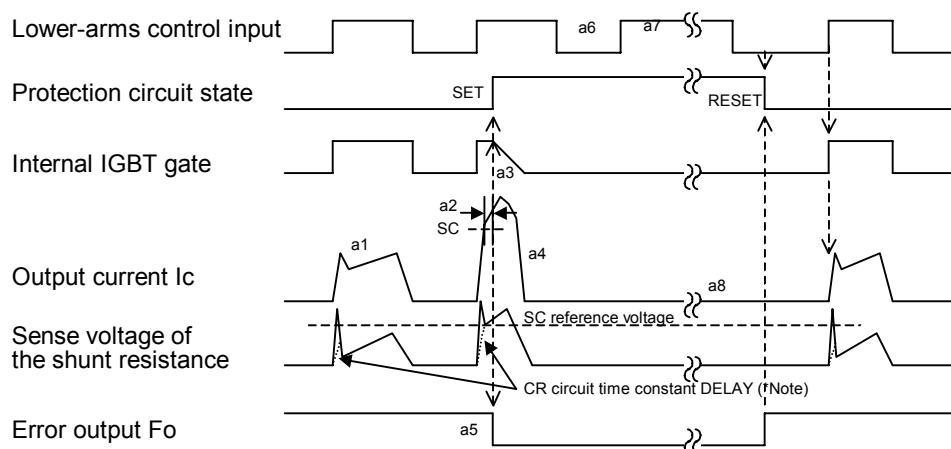
**TENTATIVE****Fig.3 DIP-IPM Internal Circuit**

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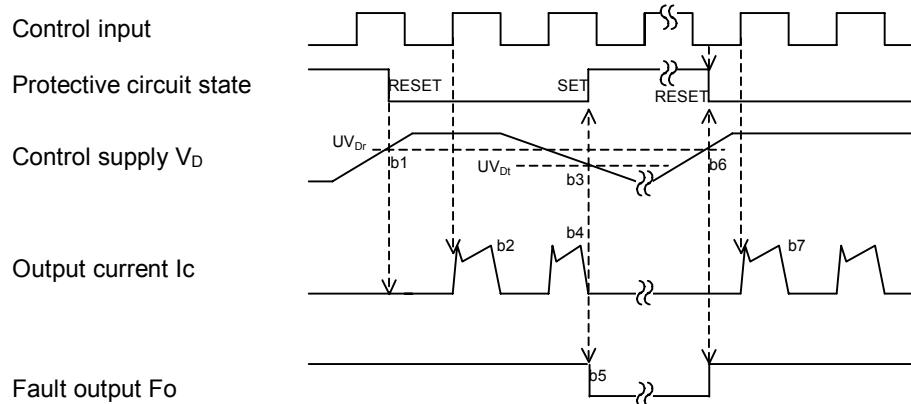
**Fig.4 Timing Charts of the DIP-IPM Protective Functions****[A] Short-Circuit Protection ( N-side only )**

(with external shunt resistor and CR filter)

- a1. Normal operation: IGBT turn on and carry current.
- a2. Short circuit current detected (SC trigger).
- a3. IGBT gate hard interrupted.
- a4. IGBT turn off.
- a5. Fo output: Fo output pulse width is determined by the external capacitance  $C_{FO}$ .
- a6. Input "L" : IGBT off.
- a7. Input "H" : IGBT on, but during the Fo output period the IGBT will not turn on.
- a8. IGBT keep in off state.

**[B] Under- Voltage Protection ( N-side,  $UV_D$  )**

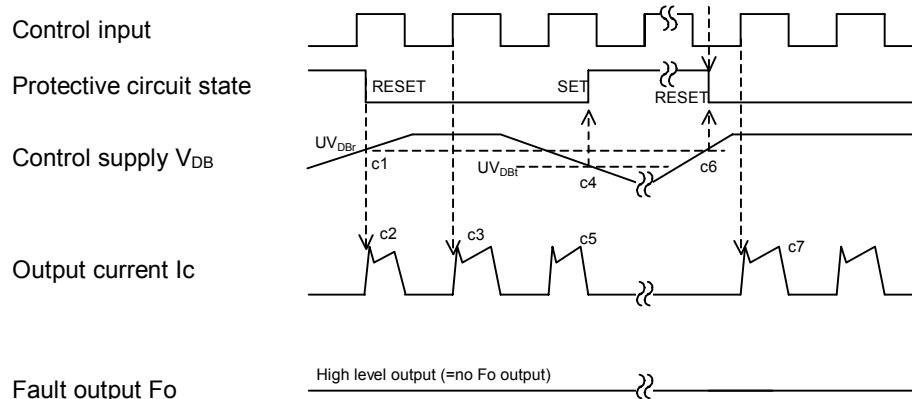
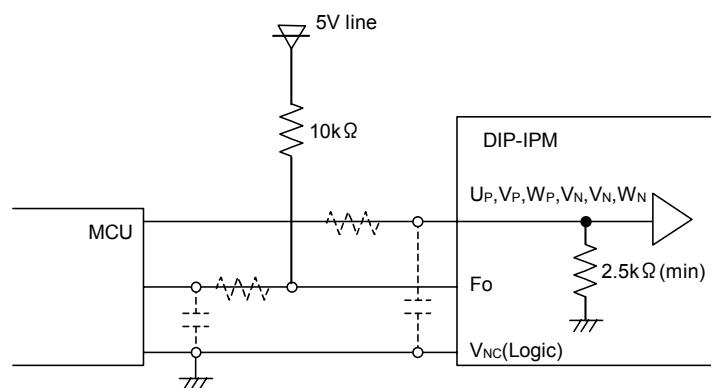
- b1. Control supply voltage rise: After the voltage level reaches  $UV_{Dr}$ , the drive circuits begin to work at the rising edge of the next input signal.
- b2. Normal operation: IGBT turn on and carry current.
- b3. Under voltage trip ( $UV_{Dt}$ ).
- b4. IGBT turn off regardless of the control input level.
- b5. Fo output.
- b6. Under voltage reset ( $UV_{Dr}$ ).
- b7. Normal operation: IGBT turn on and carry current.



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[C] Under- Voltage Protection ( P-side,  $UV_{DB}$  )

- c1. Control supply voltage rise: After the voltage reaches  $UV_{DBr}$ , the drive circuit begins to work.
- c2. Normal operation: IGBT turn on and carry current.
- c3. Under voltage trip ( $UV_{DBt}$ ).
- c4. IGBT turn off regardless of the control input level, but there is no  $Fo$  signal output.
- c5. Under voltage reset ( $UV_{DBr}$ ).
- c6. Normal operation: IGBT turn on and carry current.

**Fig.5 Recommended MCU I/O interface circuit**

Note) RC coupling at each input (parts shown dotted) may change depending on the PWM control scheme used in the application and the wiring impedance of the application's printed circuit board. The DIP-IPM input signal section integrates a 2.5kΩ (min) pull-down resistor. Therefore, when using a external filtering resistor, care must be taken to satisfy the turn-on threshold voltage requirement.

**Fig.6 Wiring method of shunt resistor circuit**